

EE-2025

Fall-99

Lecture 9

Linearity & Time-Invariance

27-Sept-99

Info: Web-CT, Lab, HW

- **RWS office hrs: Monday 1:30–3pm**
- **Survey this Week**
 - **Lab Quiz next week**
- **Lab #5: Image Processing**
 - **FM signals**
 - **Also Sampling and Reconstruction**
- **Prob Set #5 due FRIDAY**
 - **On-Line HW #4 also**

READING ASSIGNMENTS

- **This Lecture:**
 - **Chapter 5, pp. 133–152**
- **Other Reading:**
 - **Recitation: Ch. 5, pp. 127–133, 142–146**
 - **CONVOLUTION**
 - **Next Lecture: Chapter 6, start**

DEBUGGING

- **“Any Fool” can write code**
- **Debugging is the interesting part**
 - **It takes talent !!!**
- **HOWEVER,**
 - **Assume the stupid mistake is the problem**

DOMAINS: Time & Frequency

Time-Domain

- | $x[n]$
- | $x(t)$

Frequency Domain

- | Spectrum vs. f (Hz)
- | Spectrum vs. ω -hat

Move back and forth **QUICKLY**

LECTURE OBJECTIVES

BLOCK DIAGRAM REPRESENTATION

- | Components for **Hardware**
- | **Connect** Simple Filters Together to Build More Complicated Systems

GENERAL PROPERTIES of FILTERS

- | LINEARITY
- | TIME-INVARIANCE
- | ==> CONVOLUTION

LTI SYSTEMS

DIGITAL FILTERING

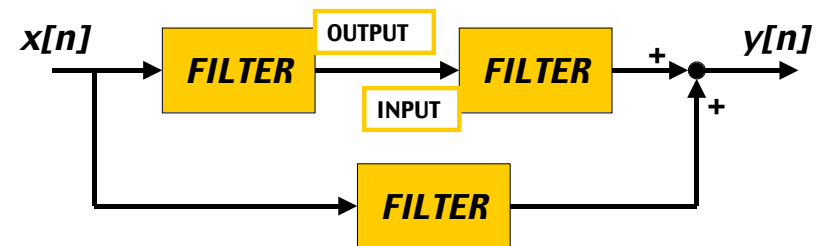


CONCENTRATE on the FILTER (DSP)

DISCRETE-TIME SIGNALS

- | FUNCTIONS of n , the “time index”
- | INPUT $x[n]$
- | OUTPUT $y[n]$

BUILDING BLOCKS



BUILD UP COMPLICATED FILTERS

- | FROM SIMPLE **MODULES**
- | Ex: FILTER **MODULE** MIGHT BE 3-pt FIR

GENERAL FIR FILTER

FILTER COEFFICIENTS $\{b_k\}$

DEFINE THE FILTER

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

For example, $\{b_k\} = \{3, -1, 2, 1\}$

$$\begin{aligned} y[n] &= \sum_{k=0}^3 b_k x[n - k] \\ &= 3x[n] - x[n - 1] + 2x[n - 2] + x[n - 3] \end{aligned}$$

MATLAB for FIR FILTER

$yy = \text{conv}(bb, xx)$

VECTOR **bb** contains Filter Coefficients

DSP-First: $yy = \text{firfilt}(bb, xx)$

FILTER COEFFICIENTS $\{b_k\}$

Conv2()
for images

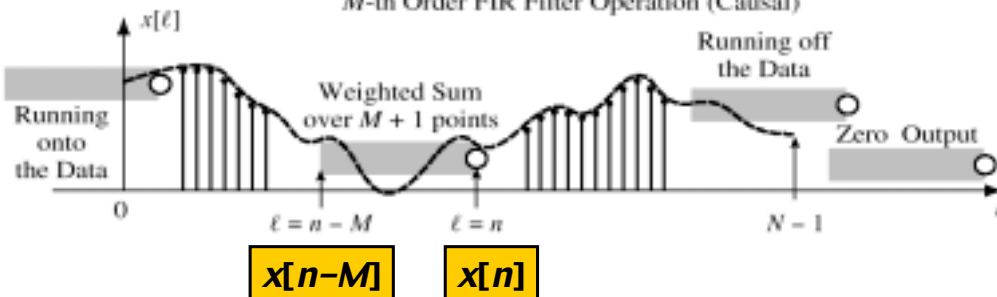
$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

GENERAL FIR FILTER

SLIDE a Length-L WINDOW over $x[n]$

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

M-th Order FIR Filter Operation (Causal)



FILTERING EXAMPLE

7-point AVERAGER $y_7[n] = \frac{1}{7} \left(\sum_{k=0}^6 x[n - k] \right)$

Removes cosine

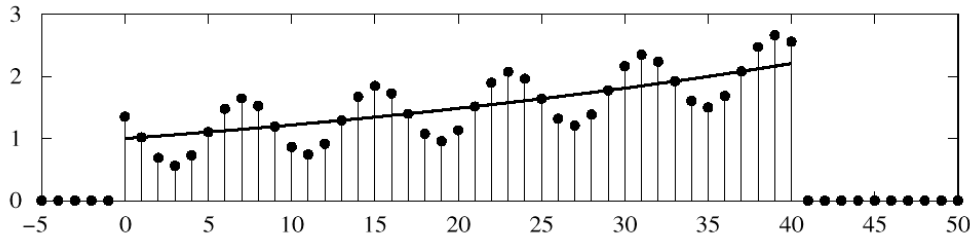
By making its amplitude (A) smaller

3-point AVERAGER $y_3[n] = \frac{1}{3} \left(\sum_{k=0}^2 x[n - k] \right)$

Changes A slightly

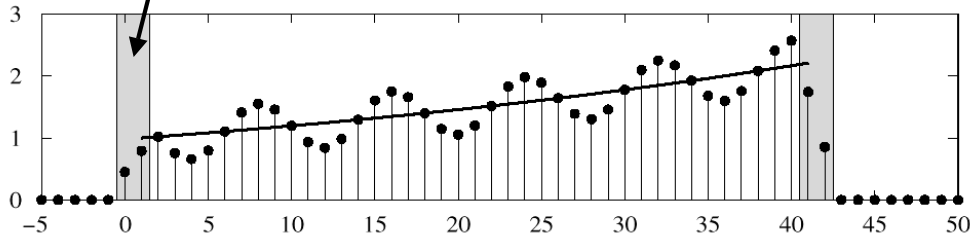
3-pt AVG EXAMPLE

Input Signal: $x[n] = (1.02)^n + \cos(2\pi n/8 + \pi/4)$ for $0 \leq n \leq 40$



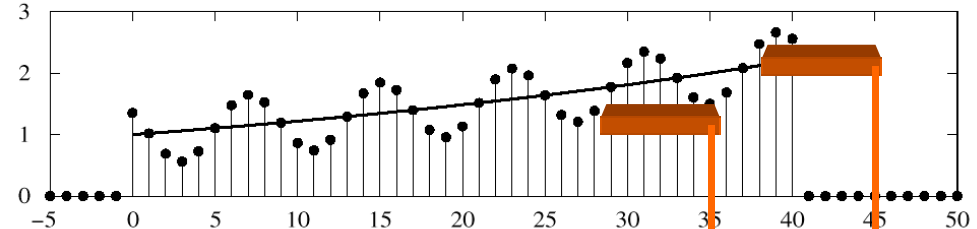
USE PAST VALUES

Output of 3-Point Running-Average Filter



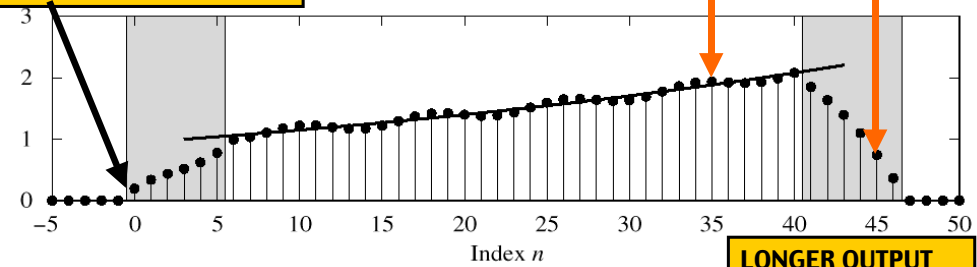
7-pt FIR EXAMPLE (AVG)

Input Signal: $x[n] = (1.02)^n + \cos(2\pi n/8 + \pi/4)$ for $0 \leq n \leq 40$



CAUSAL: Use Previous

Output of 7-Point Running-Average Filter



SPECIAL INPUT SIGNALS

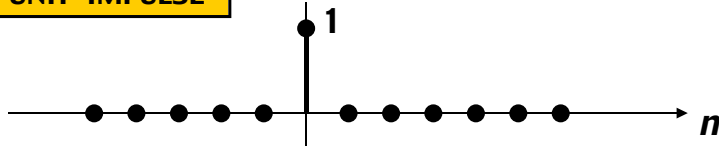
■ $x[n] = \text{SINUSOID}$

FREQUENCY RESPONSE

■ $x[n]$ has only one NON-ZERO VALUE

$$\delta[n] = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$$

UNIT-IMPULSE



UNIT IMPULSE SIGNAL $\delta[n]$

n	...	-2	-1	0	1	2	3	4	5	6	...
$\delta[n]$	0	0	0	1	0	0	0	0	0	0	0
$\delta[n-3]$	0	0	0	0	0	0	1	0	0	0	0

$\delta[n]$ is NON-ZERO When its argument is equal to ZERO

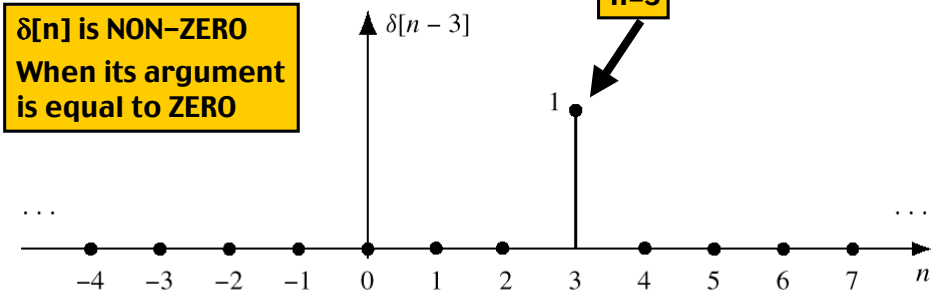
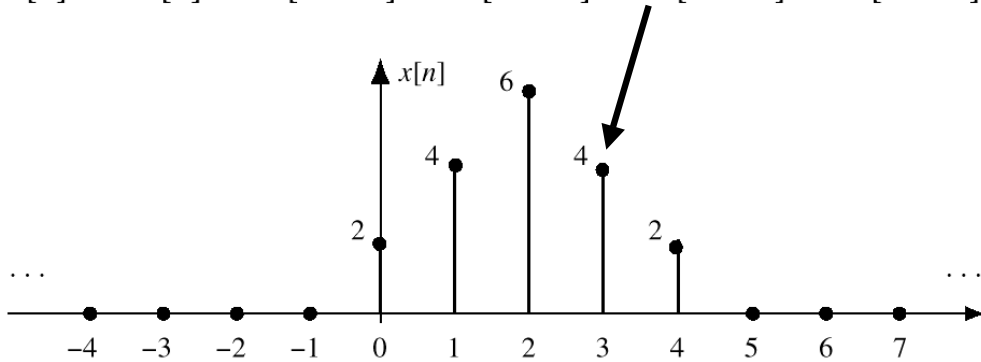


Figure 5.7 Shifted impulse sequence, $\delta[n-3]$.

MATH FORMULA for $x[n]$

Use **SHIFTED IMPULSES** to write $x[n]$

$$x[n] = 2\delta[n] + 4\delta[n - 1] + 6\delta[n - 2] + 4\delta[n - 3] + 2\delta[n - 4]$$



SUM of **SHIFTED** IMPULSES

n	...	-2	-1	0	1	2	3	4	5	6	...
$2\delta[n]$	0	0	0	2	0	0	0	0	0	0	0
$4\delta[n - 1]$	0	0	0	0	4	0	0	0	0	0	0
$6\delta[n - 2]$	0	0	0	0	0	6	0	0	0	0	0
$4\delta[n - 3]$	0	0	0	0	0	0	4	0	0	0	0
$2\delta[n - 4]$	0	0	0	0	0	0	0	2	0	0	0
$x[n]$	0	0	0	2	4	6	4	2	0	0	0

$$x[n] = \sum_k x[k]\delta[n - k] \quad \leftarrow \text{This formula ALWAYS works}$$

$$= \dots + x[-1]\delta[n + 1] + x[0]\delta[n] + x[1]\delta[n - 1] + \dots \quad (5.3.6)$$

4-pt AVERAGER

CAUSAL SYSTEM: USE PAST VALUES

$$y[n] = (x[n] + x[n-1] + x[n-2] + x[n-3])/4$$

INPUT = UNIT IMPULSE SIGNAL = $\delta[n]$

$$x[n] = \delta[n]$$

$$y[n] = 0.25\delta[n] + 0.25\delta[n-1] + 0.25\delta[n-2] + 0.25\delta[n-3]$$

OUTPUT is called "IMPULSE RESPONSE"

$$h[n] = \{\dots, 0, 0, 0.25, 0.25, 0.25, 0.25, 0, 0, \dots\}$$

FIR IMPULSE RESPONSE

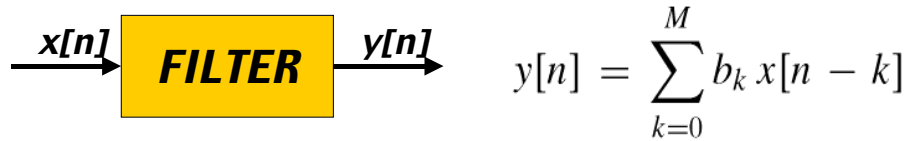
Convolution = Filter Definition

Filter Coeffs = Impulse Response

n	$n < 0$	0	1	2	3	...	M	$M + 1$	$n > M + 1$
$x[n] = \delta[n]$	0	1	0	0	0	0	0	0	0
$y[n] = h[n]$	0	b_0	b_1	b_2	b_3	...	b_M	0	0

$$y[n] = \sum_{k=0}^M h[k]x[n - k]$$

HARDWARE STRUCTURES

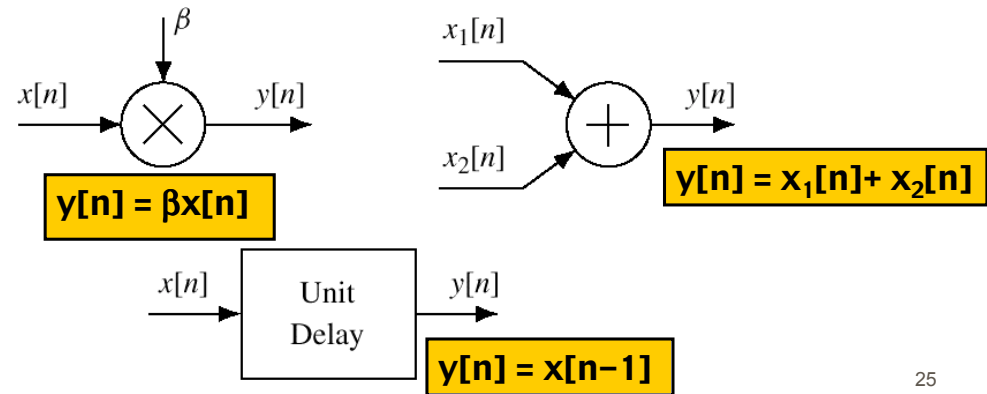


- **INTERNAL STRUCTURE** of “FILTER”
 - WHAT COMPONENTS ARE NEEDED?
 - HOW DO WE “HOOK” THEM TOGETHER?
- **SIGNAL FLOW GRAPH** NOTATION

HARDWARE ATOMS

- **Add, Multiply & Store**

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$



FIR STRUCTURE

- **Direct Form**

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

SIGNAL FLOW GRAPH

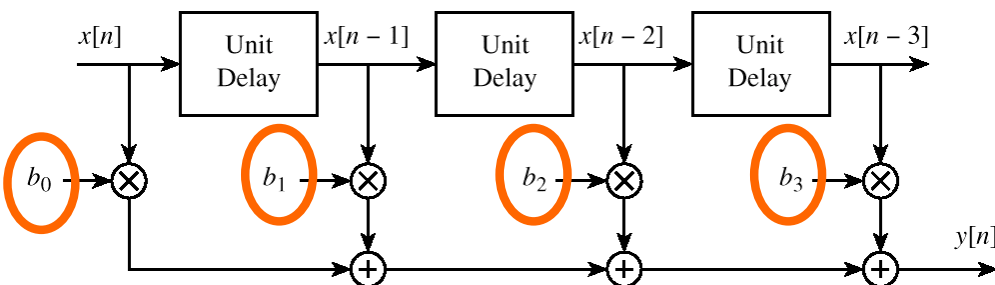
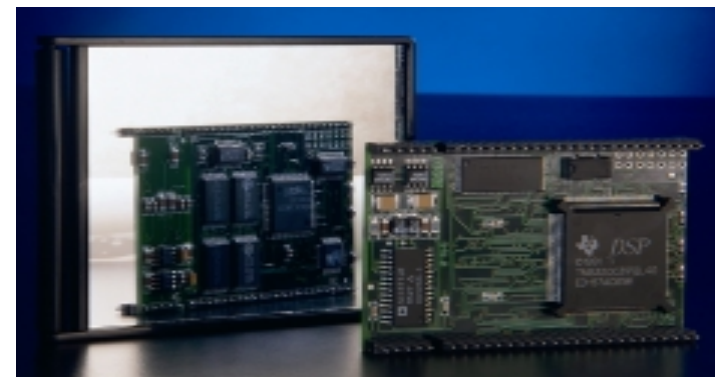
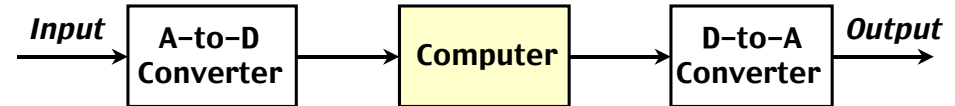
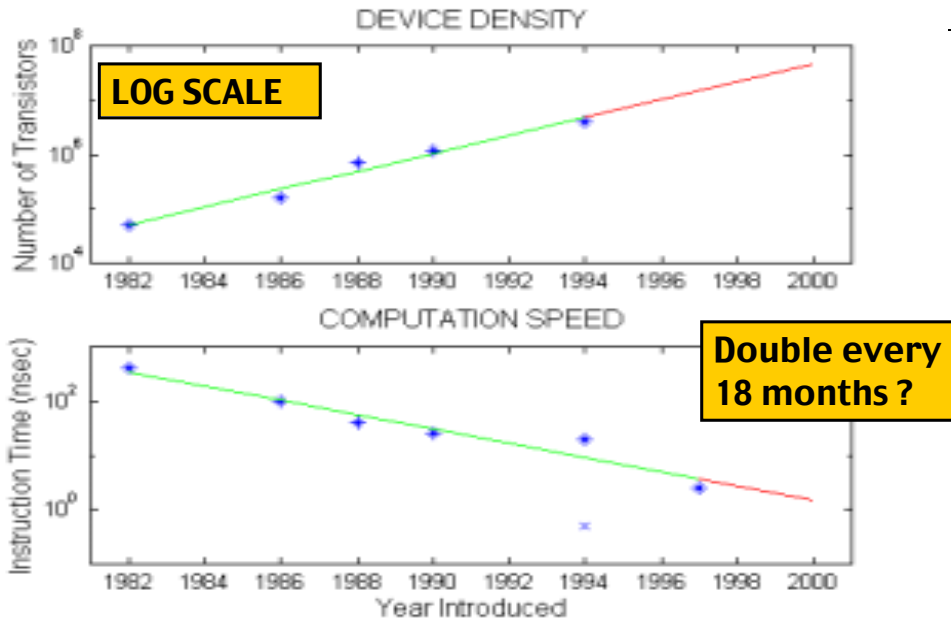


Figure 5.13 Block-diagram structure for the M th order FIR filter.

The Essence of DSP



Moore's Law for TI DSPs



SYSTEM PROPERTIES



- MATHEMATICAL DESCRIPTION
- TIME-INVARIANCE
- LINEARITY
- CAUSALITY
 - “No output prior to input”

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TIME-INVARIANCE

- **IDEA:**
 - “Time-Shifting the input will cause the **same** time-shift in the output”
- **EQUIVALENTLY,**
 - We can prove that
 - The time origin ($n=0$) is arbitrary

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TESTING Time-Invariance

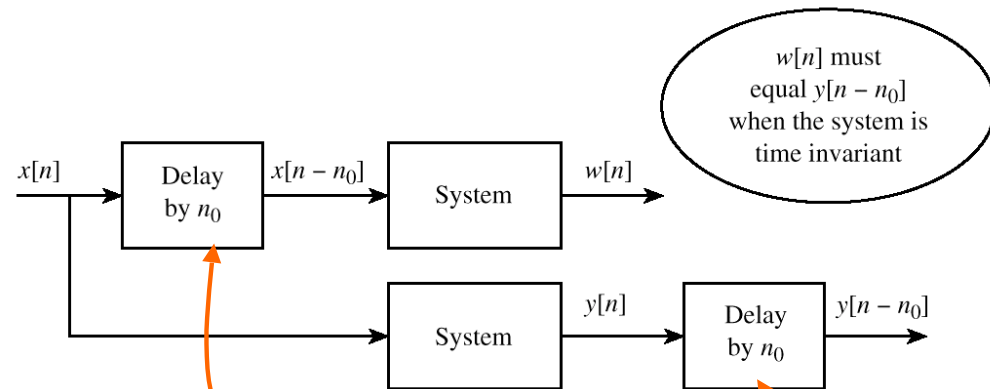


Figure 5.16 Testing time-invariance property by checking the interchange of operations.

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LINEAR SYSTEM

LINEARITY = Two Properties

SCALING

“Doubling $x[n]$ will double $y[n]$ ”

SUPERPOSITION:

“Adding two inputs gives an output that is the sum of the individual outputs”

TESTING LINEARITY

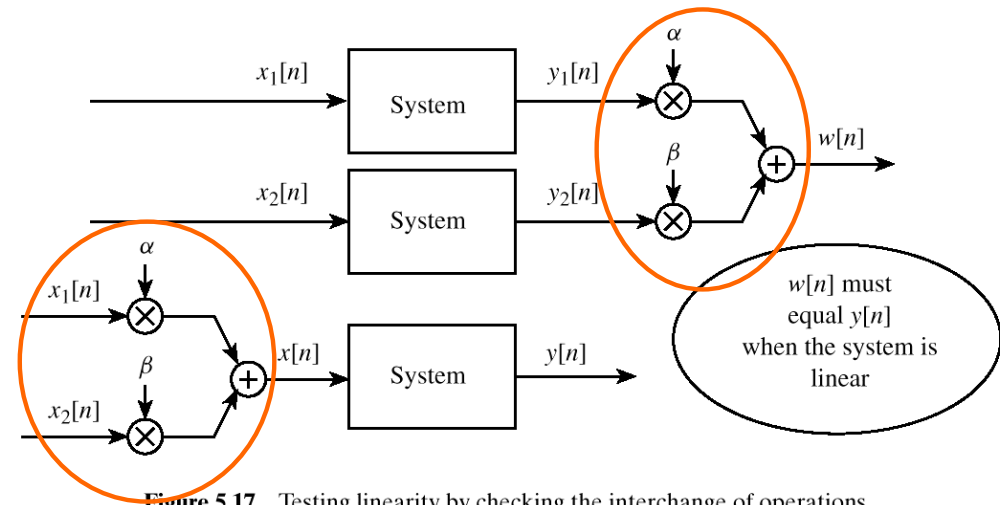


Figure 5.17 Testing linearity by checking the interchange of operations.

LTI SYSTEMS

LTI: Linear & Time-Invariant

COMPLETELY CHARACTERIZED by:

IMPULSE RESPONSE $h[n]$

CONVOLUTION: $y[n] = x[n] * h[n]$

The “rule” can be re-written as convolution

FIR Example: $h[n]$ is same as b_k

LTI: Convolution

Output = Convolution of $x[n]$ & $h[n]$

NOTATION: $y[n] = x[n] * h[n]$

Here is the FIR case:

$$y[n] = \sum_{k=0}^M h[k] x[n - k]$$

Same as b_k

FINITE LIMITS

FINITE LIMITS

CONVOLUTION Example

n	0	1	2	3	4	5	6	7	8
x[n]	2	4	6	4	2				
h[n]	3	-1	2	1					
h[0]x[n-0]	6	12	18	12	6				
h[1]x[n-1]		-2	-4	-6	-4	-2			
h[2]x[n-2]			4	8	12	8	4		
h[3]x[n-3]				2	4	6	4	2	
y[n]	6	10	18	16	18	12	8	2	