

EE-2200

Winter-99

Lecture 10

Linearity & Time-Invariance

15-Feb-99

Info: Web-CT, Lab, HW

- **Lab Quiz this week**
- **Quiz #2 on 1-March (Monday)**
- **Prob Set #5 due FRIDAY**
- **Lab #6 on Filtering**
 - **Echoes & Deconvolution**

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READING ASSIGNMENTS

- **This Lecture:**
 - **Chapter 5, pp. 133–152**
- **Other Reading:**
 - **Recitation: Ch. 5, pp. 127–133, 142–146**
 - **CONVOLUTION**
 - **Next Lecture: Chapter 6, start**

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LECTURE OBJECTIVES

- **BLOCK DIAGRAM REPRESENTATION**
 - **Components for Hardware**
 - **Connect Simple Filters Together to Build More Complicated Systems**
- **GENERAL PROPERTIES of FILTERS**
 - **LINEARITY**
 - **TIME-INVARIANCE**
 - **==> CONVOLUTION**

LTI SYSTEMS

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DIGITAL FILTERING



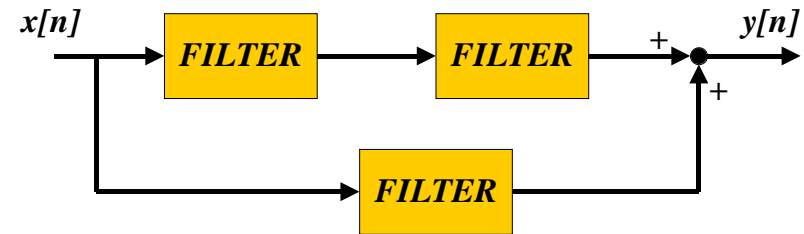
- CONCENTRATE on the FILTER (DSP)
- DISCRETE-TIME SIGNALS
 - FUNCTIONS of **n**, the “time index”
 - INPUT **x[n]**
 - OUTPUT **y[n]**

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BUILDING BLOCKS



- BUILD UP COMPLICATED FILTERS
 - FROM SIMPLE MODULES
 - Ex: FILTER MODULE MIGHT BE 3-pt FIR

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GENERAL FIR FILTER

■ FILTER COEFFICIENTS $\{b_k\}$

- DEFINE THE FILTER

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

- For example, $\{b_k\} = \{3, -1, 2, 1\}$

$$y[n] = \sum_{k=0}^3 b_k x[n - k]$$

$$= 3x[n] - x[n - 1] + 2x[n - 2] + x[n - 3]$$

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MATLAB for FIR FILTER

■ `yy = filter(bb,1,xx)`

- VECTOR **bb** contains Filter Coefficients
- DSP-First: `yy = firfilt(bb,xx)`

■ FILTER COEFFICIENTS $\{b_k\}$

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

Conv2()
for images

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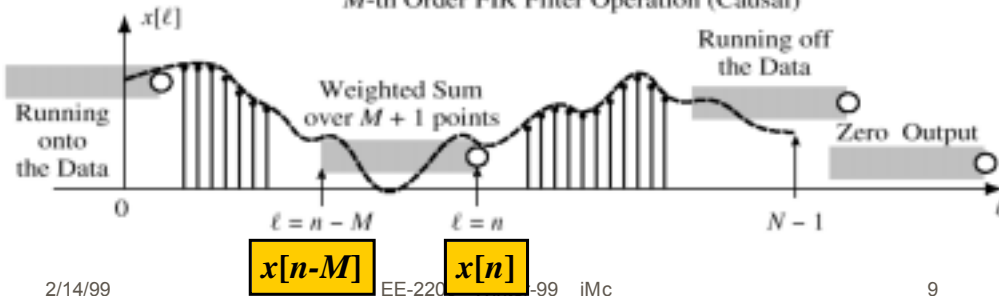
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GENERAL FIR FILTER

- SLIDE a Length-L WINDOW over $x[n]$

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

M-th Order FIR Filter Operation (Causal)



FILTERING EXAMPLE

- 7-point AVERAGER $y_7[n] = \frac{1}{7} \left(\sum_{k=0}^6 x[n - k] \right)$

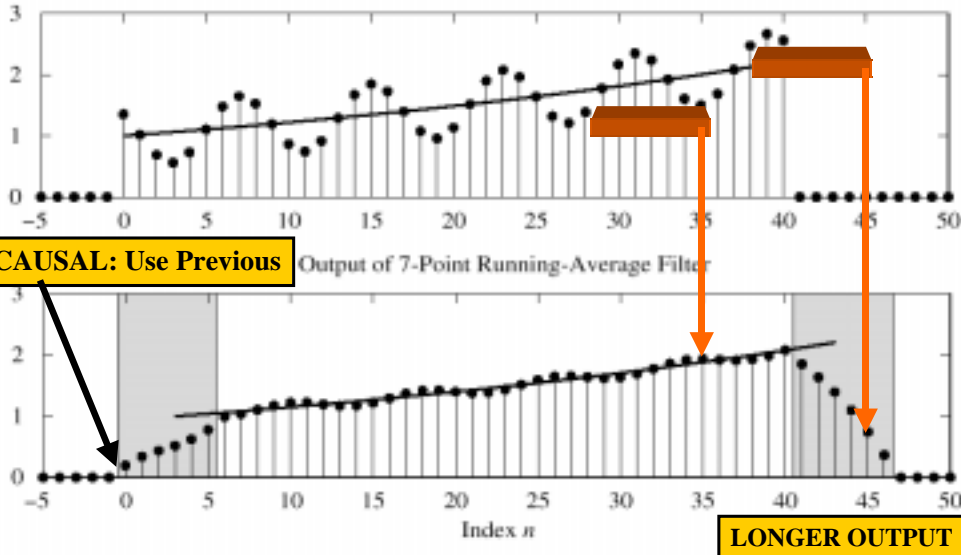
- Removes cosine

- By making its amplitude (A) smaller

- 3-point AVERAGER $y_3[n] = \frac{1}{3} \left(\sum_{k=0}^2 x[n - k] \right)$
- Changes A slightly

7-pt FIR EXAMPLE (AVG)

Input Signal: $x[n] = (1.02)^n + \cos(2\pi n/8 + \pi/4)$ for $0 \leq n \leq 40$

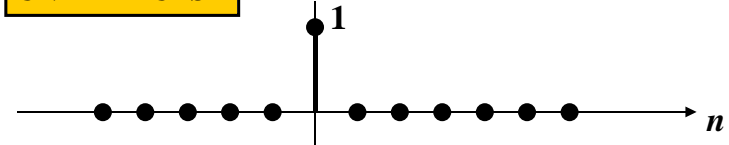


SPECIAL INPUT SIGNALS

- $x[n] = \text{SINUSOID}$ **FREQUENCY RESPONSE**
- $x[n]$ has only one **NON-ZERO VALUE**

$$\delta[n] = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$$

UNIT-IMPULSE



UNIT IMPULSE SIGNAL $\delta[n]$

n	...	-2	-1	0	1	2	3	4	5	6	...
$\delta[n]$	0	0	0	1	0	0	0	0	0	0	0
$\delta[n-3]$	0	0	0	0	0	0	1	0	0	0	0

$\delta[n]$ is NON-ZERO
When its argument
is equal to ZERO

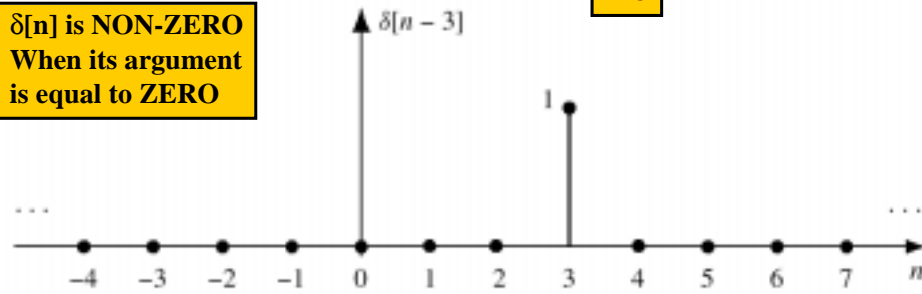
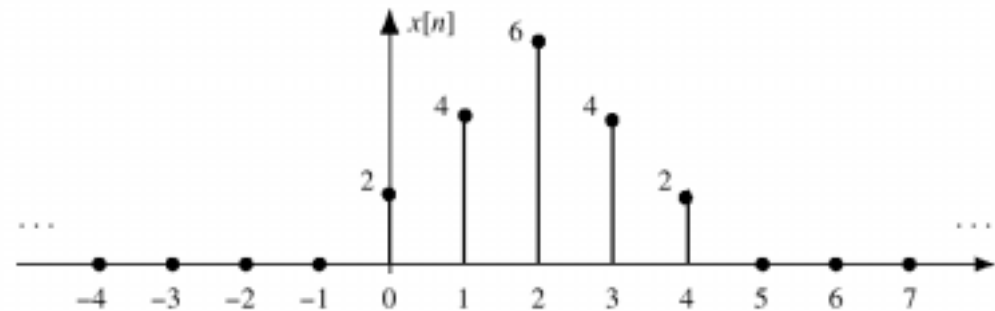


Figure 5.7 Shifted impulse sequence, $\delta[n-3]$.

MATH FORMULA for $x[n]$

Use **SHIFTED IMPULSES** to write $x[n]$

$$x[n] = 2\delta[n] + 4\delta[n-1] + 6\delta[n-2] + 4\delta[n-3] + 2\delta[n-4]$$



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SUM of **SHIFTED IMPULSES**

n	...	-2	-1	0	1	2	3	4	5	6	...
$2\delta[n]$	0	0	0	2	0	0	0	0	0	0	0
$4\delta[n-1]$	0	0	0	0	4	0	0	0	0	0	0
$6\delta[n-2]$	0	0	0	0	0	6	0	0	0	0	0
$4\delta[n-3]$	0	0	0	0	0	0	4	0	0	0	0
$2\delta[n-4]$	0	0	0	0	0	0	0	2	0	0	0
$x[n]$	0	0	0	2	4	6	4	2	0	0	0

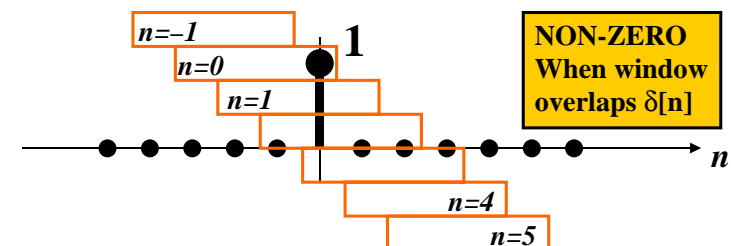
$$x[n] = \sum_k x[k]\delta[n-k]$$

← This formula ALWAYS works

$$= \dots + x[-1]\delta[n+1] + x[0]\delta[n] + x[1]\delta[n-1] + \dots \quad (5.3.6)$$

4-pt AVERAGER

- CAUSAL SYSTEM: USE PAST VALUES
- INPUT = UNIT IMPULSE SIGNAL = $\delta[n]$
- OUTPUT is called "IMPULSE RESPONSE"



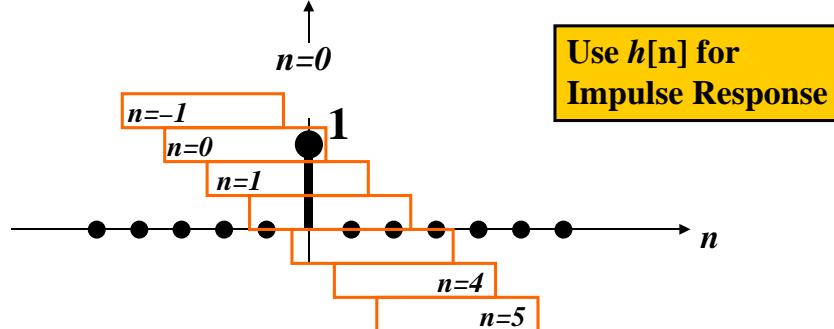
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4-pt Avg Impulse Response

- $y[n] = 0.25(x[n]+x[n-1]+x[n-2]+x[n-3])$
- “READS OUT” the FILTER COEFFICIENTS
- $y[n] = \{...,0,0,0.25,0.25,0.25,0.25,0,0,...\}$



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FIR IMPULSE RESPONSE

- Convolution = Filter Definition
- Filter Coeffs = Impulse Response

n	$n < 0$	0	1	2	3	...	M	$M+1$	$n > M+1$
$x[n] = \delta[n]$	0	1	0	0	0	0	0	0	0
$y[n] = h[n]$	0	b_0	b_1	b_2	b_3	...	b_M	0	0

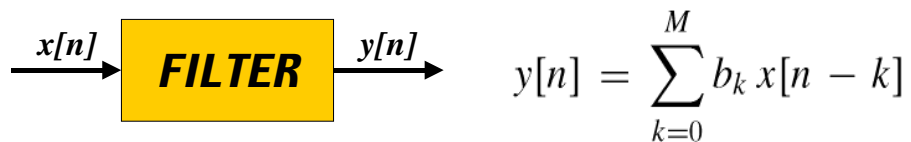
$$y[n] = \sum_{k=0}^M h[k] x[n - k]$$

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HARDWARE STRUCTURES



- INTERNAL STRUCTURE of “FILTER”
- WHAT COMPONENTS ARE NEEDED?
- HOW DO WE “HOOK” THEM TOGETHER?
- SIGNAL FLOW GRAPH NOTATION

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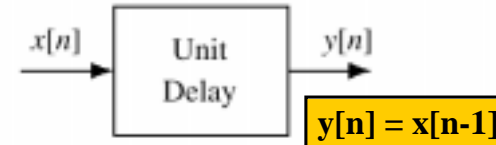
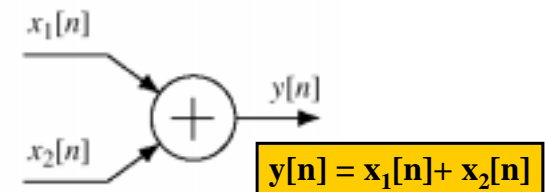
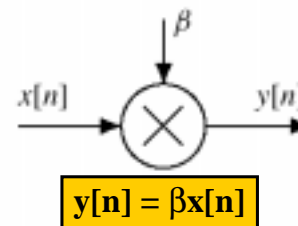
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HARDWARE ATOMS

- Add, Multiply & Store

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$



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FIR STRUCTURE

Direct Form

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

SIGNAL FLOW GRAPH

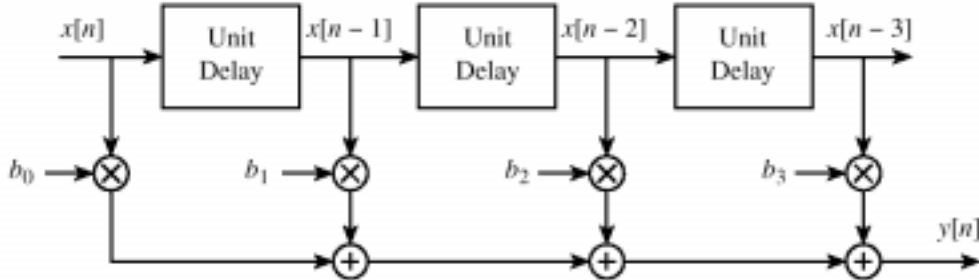


Figure 5.13 Block-diagram structure for the M th order FIR filter.

ALTERNATE STRUCTURE

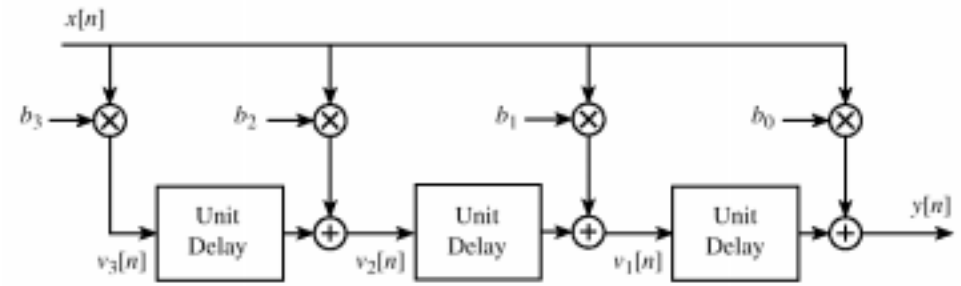


Figure 5.14 Transposed form block diagram structure for the M th order FIR filter.

$$y[n] = b_0 x[n] + v_1[n - 1]$$

$$v_1[n] = b_1 x[n] + v_2[n - 1]$$

$$v_2[n] = b_2 x[n] + v_3[n - 1]$$

$$v_3[n] = b_3 x[n]$$

DIFFERENT COMPUTATION

SYSTEM PROPERTIES



TIME-INVARIANCE

LINEARITY

CAUSALITY

“No output prior to input”

TIME-INVARIANCE

IDEA:

“Time-Shifting the input will cause the **same** time-shift in the output”

EQUIVALENTLY,

We can prove that

The time origin ($n=0$) is arbitrary

TESTING Time-Invariance

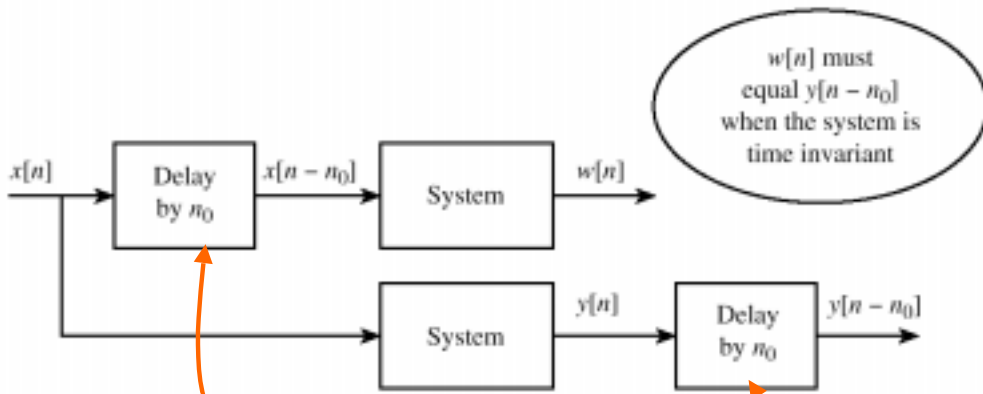


Figure 5.16 Testing time-invariance property by checking the interchange of operations.

LINEAR SYSTEM

SCALING

“Doubling $x[n]$ will double $y[n]$ ”

SUPERPOSITION:

“Adding two inputs gives an output that is the sum of the individual outputs”

TESTING LINEARITY

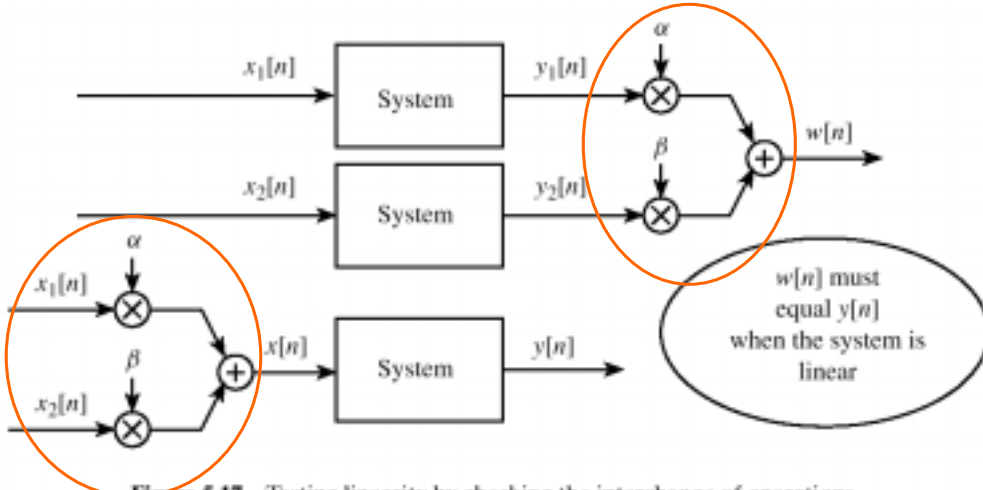


Figure 5.17 Testing linearity by checking the interchange of operations.

LTI SYSTEMS

LTI: **L**inear & **T**ime-**I**nvariant

COMPLETELY CHARACTERIZED by:

CONVOLUTION: $y[n] = x[n] * h[n]$

IMPULSE RESPONSE $h[n]$

The “rule” can be re-written as convolution

FIR Example: $h[n]$ is same as b_k

LTI: Convolution

Output = Convolution of $x[n]$ & $h[n]$

NOTATION: $y[n] = x[n]*h[n]$

Here is the FIR case:

$$y[n] = \sum_{k=0}^M h[k] x[n - k]$$

FINITE LIMITS

Same as b_k

FINITE LIMITS

CONVOLUTION Example

n	0	1	2	3	4	5	6	7	8
x[n]	2	4	6	4	2				
h[n]	3	-1	2	1					
h[0]x[n-0]	6	12	18	12	6				
h[1]x[n-1]		-2	-4	-6	-4	-2			
h[2]x[n-2]			4	8	12	8	4		
h[3]x[n-3]				2	4	6	4	2	
y[n]	6	10	18	16	18	12	8	2	

CASCADE SYSTEMS

Does the order of S_1 & S_2 matter?

NO, LTI SYSTEMS can be rearranged !!!

WHAT ARE THE FILTER COEFFS? $\{b_k\}$

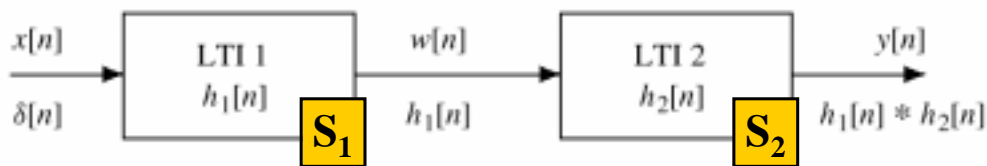


Figure 5.19 A Cascade of Two LTI Systems.

CASCADE EQUIVALENT

Find "overall" $h[n]$ for a cascade ?

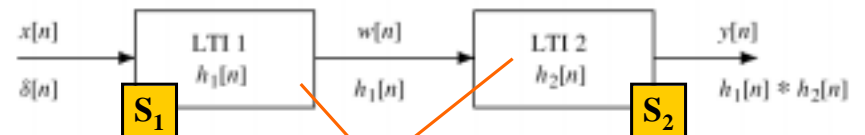


Figure 5.19 A Cascade of Two LTI Systems.

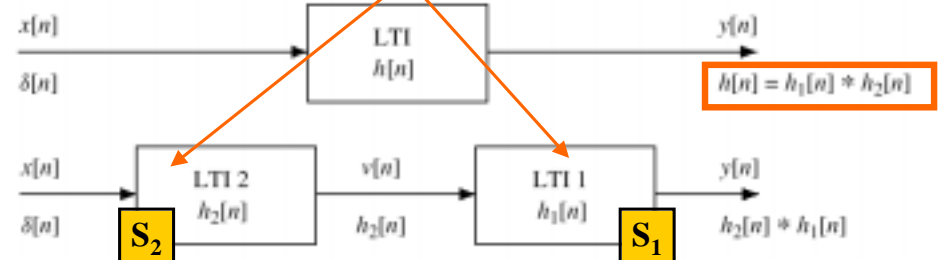


Figure 5.20 Switching the order of cascaded LTI systems.